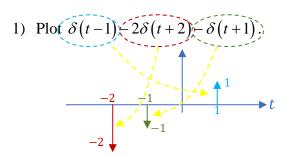
EES 351: In-Class Exercise # 1 - Sol

Instructions

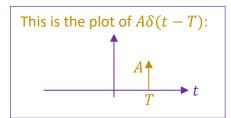
- Work alone or in a group of no more than three students. For group work, the group cannot be the same as any of your 1.
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 - (You may add the IDs of other members, exercise #, or other information as well.)
- (b) Hardcopy submission Do not panic

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Date: 21 / 8 / 2020

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- 2) Evaluate the following integrals:
 - a) $\int \delta(t)dt = 1$ because $0 \in (-\pi, \pi)$.

Remark: As discussed in class, writing the limits of integration in this form is ambiguous when dealing with δ -function because we don't know whether the boundary points are included in the integration or not. However, this does not affect our answer in this part because 0 will still be within the interval of integration even without the boundary points.

b)
$$\int_{\pi}^{2\pi} \delta(t-3)dt = 0$$

$$\int_{\pi}^{\pi} c = 3$$

$$A = [\pi, 2\pi] \} c \notin A$$

$$g(t) \equiv 1$$

c)
$$\int_{\pi}^{2\pi} e^{j\frac{\pi}{2}t} \delta(t)dt = 0$$

$$\int_{\pi}^{2\pi} c = 0$$

$$A = [\pi, 2\pi] \} c \notin A$$

Remark: As discussed in class, writing the limits of integration in this form is ambiguous when dealing with δ function because we don't know whether the boundary points are included in the integration or not. However, this does not affect our answer in this part because c will still be outside A even when we also include the two boundary points.

$$\int_{A} \delta(t) dt = \begin{cases} 1, & 0 \in A \\ 0, & 0 \notin A \end{cases}$$

(Extended) Sifting Property:

$$\int_{A} g(t)\delta(t-c) dt = \begin{cases} g(c), & c \in A \\ 0, & c \notin A \end{cases}$$

$$g(t) = e^{j\frac{\pi}{2}t}$$

$$d) \int_{0}^{2\pi} e^{j\frac{\pi}{2}t} \delta(t-3) dt = g(c) = e^{j\frac{\pi}{2}3} = -j$$

$$c = 3$$

$$A = (0,2\pi)$$

$$c \in A$$
Similar remark to part (a)
$$g(t) = e^{j\frac{\pi}{2}t}$$

= 1

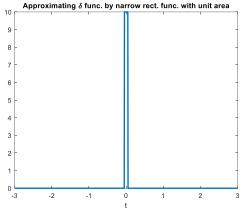
e)
$$\int_{0}^{\infty} \delta(t-1) - 2\delta(t+2) - \delta(t+1)dt = \int_{0}^{\infty} \delta(t-1) dt - 2\int_{0}^{\infty} \delta(t+2) dt + \int_{0}^{\infty} \delta(t+1) dt - \frac{1}{2} \int_{0}^{\infty} \delta(t+1) dt + \int_{0}^{\infty} \delta($$

f) (optional)
$$\int_{-\infty}^{\infty} \delta(t^2 - 2t) dt$$

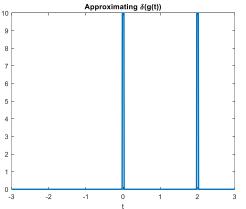
In the optional question, we would like to calculate $\int_{-\infty}^{\infty} \delta(g(t)) dt$ where $g(t) = t^2 - 2t$.

Here are some hints.

Let's look at the limiting approximation of $\delta(t)$. Consider a rectangular function centered at origin whose width is ε . To be a delta function, we need the area = 1; therefore, the height must be $\frac{1}{\varepsilon}$. In the MATLAB plot below, we use $\varepsilon = 0.1$. As expected, there is a "spike" at t = 0.



Now, let's try to plot $\delta(g(t))$ where $g(t) = t^2 - 2t$.





Note that there are two "spikes" at t = 0 and t = 2. This is expected because we know that the spikes will show up when the argument of $\delta(\cdot)$ is 0. Here, $g(t) = t^2 - 2t$ is zero at t = 0 and t = 2.

The area under the graph above approximates $\int_{-\infty}^{\infty} \delta(g(t)) dt$. This area is the sum of the areas under

the two spikes. Note, however, that the area under each spike is not one anymore; the spikes seem to be narrower. How can we find their areas? (Furthermore, can we eliminate MATLAB from this calculation?)

Date: 26 / 8 / 2020

b. $g(t) = 3\cos(4\pi t)$

1.5

 $A\cos(2\pi f_0 t)$

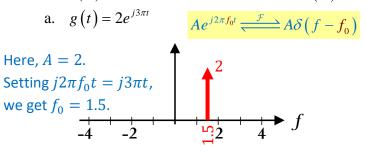
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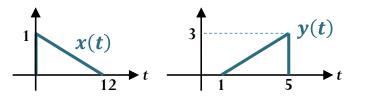
Instructions

- - (b) Hardcopy submission5. Do not panic.
- 1. [ENRPr] Consider each g(t) defined below.

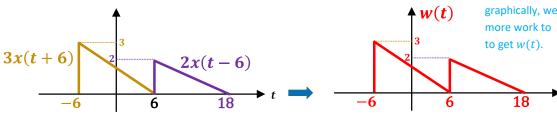
Let G(f) be its Fourier transform. Plot G(f) from f = -4 to f = 4 Hz.



2. [ENRPr] Signals x(t) and y(t) are plotted below.



a) Plot the signal
$$w(t) = 2x(t-6) + 3x(t+6)$$
.



Remark: There is no nonzero overlapping part between 3x(t + 6) and 2x(t - 6). Therefore, graphically, we don't need to do any more work to "add" the two graphs to get w(t).

ID (last 3 digits)

Don't forget to

scale the size

function by $\frac{1}{2}$.

of each δ -

1.5

 $\frac{A}{2}\delta(f-(-f_0))+\frac{A}{2}\delta(f-f_0)$

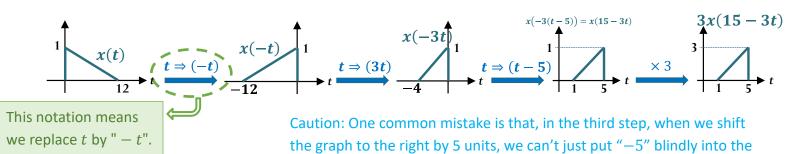
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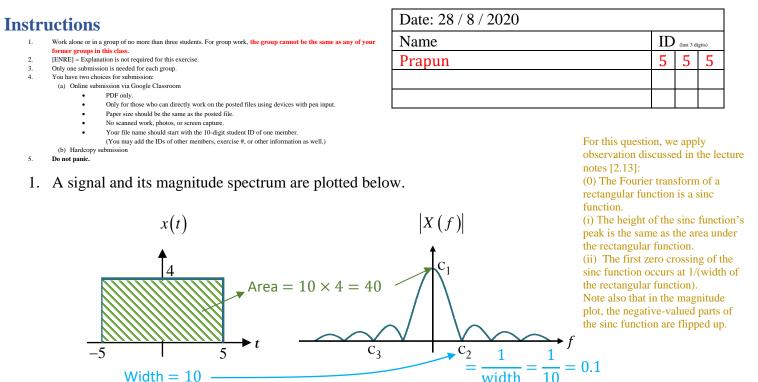
b) Suppose $y(t) = c_1 x (c_2 t + c_3)$. Find the values of the constants c_1, c_2 , and c_3 .

$$c_1 = 3$$
 , $c_2 = -3$, $c_3 = 15$

expression and get x(-3t-5); we need to replace t by t-5.



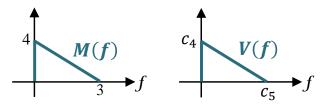
EES 351: In-Class Exercise # 3 - Sol



Find the values of the constants (corresponding to some zeroes and the peak value) shown in the plots.

 $c_1 = 40$, $c_2 = 0.1$, $c_3 = -2c_2 = -0.2$

2. Consider a signal m(t) and another signal v(t) = m(5t). Their corresponding Fourier transforms are plotted below.



Caution: The relationship between the two signals is given in the time domain. However, the plots are given in the frequency domain.

Find the values of the constants in the plot of V(f):

$$c_4 = \frac{4}{5} = 0.8, c_5 = 15$$

For v(t) = m(5t), by the scale-change theorem [2.32 eq. (21)], we have

$$V(f) = \frac{1}{|5|} M\left(\frac{f}{5}\right) = \frac{1}{5} M\left(\frac{f}{5}\right).$$

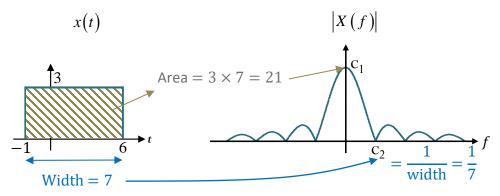
In the previous exercise, we worked on time manipulation. Note that, back then, "time" was our dummy variable. Here, it's the frequency f. We can get $M\left(\frac{f}{5}\right)$ from M(f) by replacing f by $\frac{f}{5}$; therefore, graphically, this is a horizontal expansion by a factor of 5. This implies $c_5 = 5 \times 3 = 15$.

Finally, the $\frac{1}{5}$ in the front simply scales the height of graph by a factor of $\frac{1}{5}$. This implies $c_4 = \frac{1}{5} \times 4 = \frac{4}{5} = 0.8$.

Instructions

3. 4.

- 1. Work alone or in a group of no more than three students. For group work, the group cannot be the same as any of your
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 - You may add the IDs of other members, exercise #, or other information as well.)
 rdcopy submission
- (b) Hardcopy 5. **Do not panic.**
- 1. A signal and its magnitude spectrum are plotted below.



Find the values of the constants (corresponding to some zeroes and the peak value) shown in the plots.

$$c_1 = \underline{21}, c_2 = \underline{\frac{1}{7}}$$

This problem is similar to the one we have worked on in the previous exercise. However, the rectangular function is not centered at t = 0; it is time-shifted. From the time-shift property (2.31) of Fourier transform, we know that the magnitude spectrum plot won't be affected by this time-shifting. So, we can still use 2.13. In particular,

(0) The Fourier transform of a rectangular function is a sinc function.

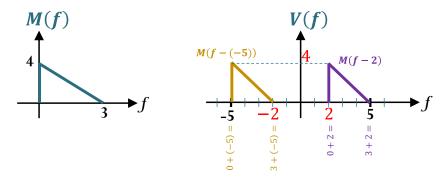
(i) The height of the sinc function's peak is the same as the area under the rectangular function.

(ii) The first zero crossing of the sinc function occurs at 1/(width of the rectangular function).

2. Consider a signal m(t). Its Fourier transform M(f) is plotted below.

Let
$$v(t) = e^{-j10\pi t} m(t) + e^{j4\pi t} m(t) = e^{j2\pi(-5)t} m(t) + e^{j2\pi(2)t} m(t)$$

Plot V(f) in the corresponding space below.



Note that the plots of M(f - (-5)) and M(f - 2) do not have any (nonzero) overlapping part, so the plot of their sum is the same as what we already have here.

From the frequency-shift property of Fourier transform, we know that $e^{j2\pi f_0 t}m(t) \xrightarrow{\mathcal{F}} M(f - f_0)$.

In particular, $e^{j2\pi(-5)t}m(t) \xrightarrow{\mathcal{F}} M(f-(-5))$ and $e^{j2\pi(2)t}m(t) \xrightarrow{\mathcal{F}} M(f-2)$.

Therefore, $e^{j2\pi(-5)t}m(t) + e^{j2\pi(2)t}m(t) \xrightarrow{\mathcal{F}} M(f-(-5)) + M(f-2).$

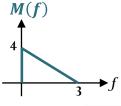
Date: 2 / 9 / 2020			
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Instructions

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- (b) Hardcopy submission 5 Do not panic.
- 1. Consider a signal m(t). Its Fourier transform M(f) is plotted below.



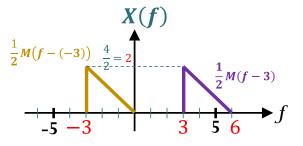
a. Let $x(t) = \cos(6\pi t)m(t)$.

Plot X(f) in the corresponding space below.

Recall that

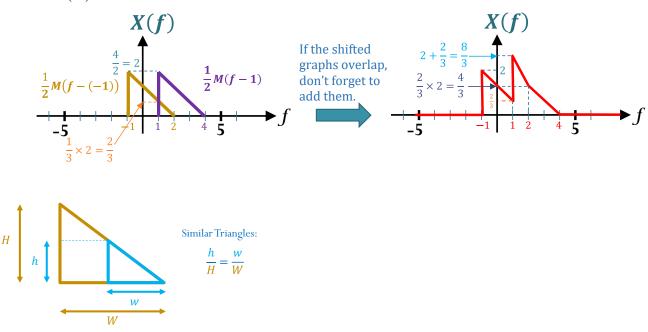
ID (last 3 digits)

$$g(t)\cos(2\pi(f_c)t) \stackrel{F}{\rightleftharpoons} \frac{1}{2}G(f-f_c) + \frac{1}{2}G(f-(-f_c)).$$



b. Let $x(t) = \cos(2\pi t)m(t)$.

Plot X(f) in the corresponding space below.



Date: 9 / 9 / 2020	
Name	

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(b) Hardcopy submission Do not panic 5

In this problem, we have three "devices".

- is a "square" device. As the name suggests, its output is created by squaring its input in the time domain.
- $H_1(f)$ is an LTI device whose <u>frequency response</u> is $H_1(f) = \begin{cases} 1, & |f| < 315, \\ 0, & \text{otherwise.} \end{cases}$
- $H_2(f)$ is an LTI device whose <u>frequency response</u> is $H_2(f) = \begin{cases} 1, & |f| > 315, \\ 0, & \text{otherwise.} \end{cases}$

Find the output y(t) for each of the systems below.

$$j2\pi f_0 t = 351\pi t \implies f_0 = \frac{351\pi}{2} = 175.5$$
(a) $x(t) = e^{j351\pi t} \longrightarrow H_1(f) \longrightarrow y(t)$
 $H_1(175.5) = 1 \text{ because } |175.5| < 315.$

$$y(t) = H_1(f_0)e^{j2\pi f_0 t} = H_1(175.5)e^{j2\pi(175.5)t} = 1e^{j351\pi t} = e^{j351\pi t}$$

(b)
$$x(t) = \cos(351\pi t) \longrightarrow H_1(f) \longrightarrow y(t)$$

$$y(t) = H_1(f_0)\cos(2\pi f_0 t) = H_1(175.5)\cos(2\pi (175.5)t) = \cos(351\pi t)$$

$$\Rightarrow \text{Recall that}$$

$$\cos(2\pi f_0 t) \rightarrow H(f) \rightarrow \frac{1}{2}H(f_0)e^{j2\pi f_0 t} + \frac{1}{2}H(-f_0)e^{-j2\pi f_0 t} + \frac{1}{2}$$

(c)
$$x(t) = \cos(351\pi t) \longrightarrow H_2(f) \longrightarrow y(t)$$

 $y(t) = H_2(f_0) \cos(2\pi f_0 t) = H_2(175.5) \cos(2\pi (175.5)t) = 0$
 $H_2(175.5) = 0$ because $|175.5| \ge 315$.

(d)
$$x(t) = \cos(351\pi t) \longrightarrow (\cdot)^2 \xrightarrow{x^2(t)} H_1(f) \longrightarrow y(t)$$
 whose freq. is 0
(e) $x^2(t) = \cos^2(351\pi t) = \left(\frac{e^{j351\pi t} + e^{-j351\pi t}}{2}\right)^2 = \frac{1}{4}e^{j2\pi(351)t} + \left(\frac{1}{2}\right) + \frac{1}{4}e^{j2\pi(-351)t}$ So, $x^2(t) = \frac{1}{4}H_1(351)e^{j2\pi(351)t} + \frac{1}{2}H_1(0) + \frac{1}{4}H_1(-351)e^{j2\pi(-351)t} = \frac{1}{2}$ we can

So, $x^2(t)$ is simply a linear combination of complexexponential functions. Therefore, we can apply our \star to each term.

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Recall that

One can view the constant $\frac{1}{2}$ as a

 $\frac{1}{2}e^{j2\pi(0)t}$

complex-expo. function

H(f)

 $\rightarrow H(f_0)e^{j2\pi f_0 t}$

Instructions

4

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(b) Hardcopy submission Do not panic.

1. Consider a channel with multipath propagation. Its impulse response is of the form

$$h(t) = \sum_{k=1}^{\nu} \beta_k \delta(t - \tau_k)$$

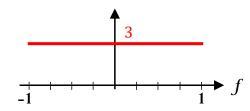
a. Suppose v = 2, $\beta_1 = \beta_2 = 3$, $\tau_1 = 2$, $\tau_2 = 5$.

For each of the following channel input x(t), find the corresponding channel output y(t). Note that the output should be of the form $y(t) = A\cos(2\pi f_0 t + \theta_0)$ for some constants A, f_0 , and θ_0 where θ_0 is in degrees.

Channel input	Channel output]
$x(t) = \cos(\pi t)$	$= \frac{3\cos(\theta - \cos(\theta + 2\times 2\pi))}{\cos(\theta - \pi)} - 3\cos(\pi t) \qquad \cos(\theta - \pi) = \cos(\theta)$	π period of cosine: cos(θ) = cos(θ + n2π) or any integer <i>n</i> . and freq. are OK here.
$x(t) = \cos\left(\frac{\pi}{2}t\right)$ Conversion	y(t) = 3x(t-2) + 3x(t-5) = $3\cos\left(\frac{\pi}{2}(t-2)\right) + 3\cos\left(\frac{\pi}{2}(t-5)\right)$ = $3\cos\left(\frac{\pi}{2}t-\pi\right) + 3\cos\left(\frac{\pi}{2}t-\frac{5\pi}{2}\right)$ sion to phasor form $\Leftrightarrow 3 \angle -180^\circ + 3 \angle -90^\circ = 3\sqrt{2}\angle -135^\circ$ eack to time domain $\Leftrightarrow 3\sqrt{2}\cos\left(\frac{\pi}{2}t\right) - 135^\circ$	We can use phasor representation to combine sinusoids with the same frequency.

Note that the $2\pi f_0 t$ part of the cosine should be the same. Here, it is $\frac{\pi}{2}t$.

b. Suppose v = 1, $\beta_1 = 3$, $\tau_1 = 2$. Plot |H(f)| from f = -1 to f = 1 Hz.



When $\nu = 1$, we have $h(t) = \beta_1 \delta(t - \tau_1)$. With the provided values, we have $h(t) = 3\delta(t - 2)$. Therefore, $H(f) = 3e^{-j2\pi(2)f}$ and $|H(f)| = 3|e^{-j4\pi f}| = 3 \times 1 = 3$.

Note that this is a distortionless channel. So, the magnitude spectrum should be flat.

Date: 16 / 9 / 2020

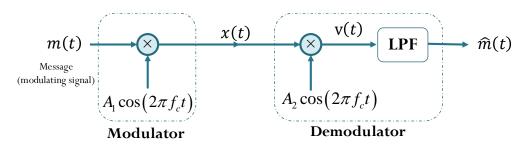
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Date: 18 / 9 / 2020

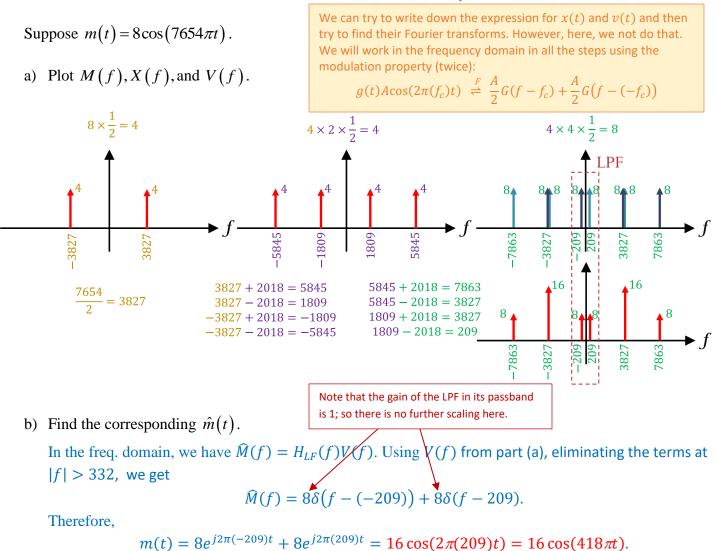
Instructions

3. 4.

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- (b) Hardcopy 5. **Do not panic.**
- 1. Consider a DSB-SC modem with no channel impairment shown below.



Let $A_1 = 2$, $A_2 = 4$, and $f_c = 2018$ Hz. Suppose LPF has $H_{LP}(f) = \begin{cases} 1, & |f| \le 332, \\ 0, & \text{otherwise.} \end{cases}$



Name	ID (last 3 digits)				

Alternatively, we can work in the time domain directly.

We will need to use a trigonometric identity that we have proved in our HW:

$$\cos(A)\cos(b) = \frac{1}{2}(\cos(A+B) + \cos(A-B)).$$

Here, we have

$$v(t) = x(t) \times 4\cos(2\pi(2018)t) = (m(t) \times 2\cos(2\pi(2018)t)) \times 4\cos(2\pi(2018)t)$$

= m(t) × 8 cos²(2\pi(2018)t) = m(t) × 4(1 + cos(2\pi(2 × 2018)t))
= 4m(t) + 4m(t) cos(2\pi(2 × 2018)t).
ituting m(t) = 8 cos(7654\pi t) = 8 cos(2\pi(3827)t), we get

Substituting $m(t) = 8\cos(7654\pi t) = 8\cos(2\pi(3827)t)$, we get $w(t) = 32\cos(2\pi(3827)t) + 32\cos(2\pi(3827)t)\cos(2\pi(4036)t)$

$$v(t) = 32\cos(2\pi(3827)t) + 32\cos(2\pi(3827)t)\cos(2\pi(4036)t)$$

 $= 32\cos(2\pi(3827)t) + 16\cos(2\pi(209)t) + 16\cos(2\pi(7863)t).$

After the LPF, only sinusoid with freq. not exceeding 332 passes through. Therefore,

 $m(t) = 16\cos(2\pi(209)t) = 16\cos(418\pi t).$

Instructions

4

- . Work alone or in a group of no more than three students. For group work, the group cannot be the same as any of your former groups in this class.
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- (b) Hardcopy5. Do not panic.
- 1. Consider the "modulator" shown below. Note that the first operation is a summation (not multiplication).

$$m(t) \xrightarrow{\qquad } (\cdot)^{2} \xrightarrow{\nu(t)} H_{\rm BP}(f) \xrightarrow{} x(t)$$
$$A_{c} \cos(2\pi f_{c} t)$$

 $(\cdot)^2$ is a "square" device; its output is created by squaring its input in the <u>time</u> domain.

 $|H_{\rm BP}(f)|$ is an LTI device whose <u>frequency response</u> is

$$H_{\rm BP}(f) = \begin{cases} 1, & |f - f_c| \le 332, \\ 1, & |f + f_c| \le 332, \\ 0, & \text{otherwise.} \end{cases}$$

Let $A_c = 2$, $f_c = 2018$, and

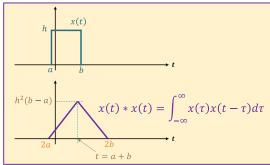
$$M(f) = \begin{cases} 1, & |f| \le 54, \\ 0, & \text{otherwise.} \end{cases}$$

a. Plot the corresponding X(f).

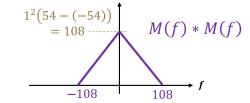
 $v(t) = (m(t) + 2\cos(2\pi(2018)t))^2 = m^2(t) + 4\cos^2(2\pi(2018)t) + 4m(t)\cos(2\pi(2018)t).$ v(t) is input into a band-pass filter. So, we should look at V(f).

There are three terms in v(t). Let's analyze them in the frequency domain separately.

(1) By the convolution-in-frequency property: $m^2(t) = m(t) \times m(t) \stackrel{F}{\rightleftharpoons} M(f) * M(f)$.



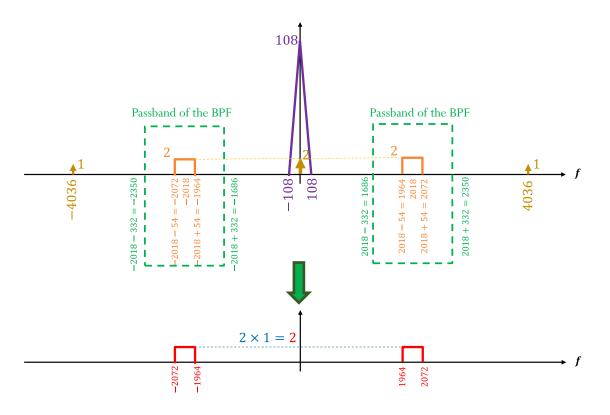
In lecture, we have seen how to do self-convolution of a rectangular function. In this problem, a = -54, b = 54, h = 1, and the dummy variable is f instead of t. This gives



(Actually, the actual shape of the convolution is not as important as the fact that the result is band-limited to 128 Hz.

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(2) From $\cos^2 \theta = \frac{1}{2}(1 + \cos(2\theta))$, we have $4\cos^2(2\pi(2018)t) = 2 + 2\cos(2\pi(4036)t).$ (3) Fourier transform of the term $4m(t)\cos(2\pi(2018)t)$ can be found by our modulation formula: $g(t)A\cos(2\pi(f_c)t) \stackrel{F}{\rightleftharpoons} \frac{A}{2}G(f - f_c) + \frac{A}{2}G(f - (-f_c)).$

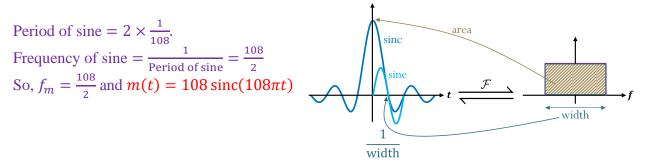


b. (Optional) Find m(t) and x(t).

M(f) is a rectangular function in the frequency domain. So, m(t) is a sinc function in the time domain. The area of the rectangular function is $1 \times 54 - (-54) = 108$. So, $m(t) = 108 \operatorname{sinc}(2\pi f_m t),$

for some f_m .

The width of the rectangular function is 54 - (-54) = 108. So, the first zero-crossing of the sinc function should occur at $t = \frac{1}{108}$. From sinc $(\theta) = \frac{\sin(\theta)}{\theta}$, we see that the zero-crossing of sinc (θ) are the same as the zero-crossings of sin (θ) (except when $\theta = 0$).



Back in the analysis of v(t), we saw that $4m(t)\cos(2\pi(2018)t)$ passes through the filter unchanged. So, $x(t) = 4m(t)\cos(2\pi(2018)t) = 432\sin(108\pi t)\cos(4026\pi t)$.