#### Instructions

- 1. Separate into groups of no more than three persons. The group cannot be the same as any of your former groups. Only one submission is needed for each group.
- Write down all the steps that you have done to obtain your answers. You may not get full credit even when your answer is correct without showing how you get your answer.
- 3. Do not panic.
- 1. Assume GF(2). Calculate the following quantities:
  - a. 1⊕1 **=**
  - b.  $1 \oplus 1 \oplus 1 = (1 \oplus 1) \oplus 1 = 0 \oplus 1 = 1$
  - c. 1·0 **≂** O
  - d. 1 ⋅ 0 ⋅ 1 **=**
  - e.  $\begin{bmatrix} 1 & 1 & 1 \end{bmatrix} \oplus \begin{bmatrix} 1 & 0 & 1 \end{bmatrix} = \begin{bmatrix} 1 \textcircled{0} & 1 & 1 \textcircled{0} & 1 \end{bmatrix} = \begin{bmatrix} 0 & 1 & 0 \end{bmatrix}$
  - f.  $\begin{bmatrix} 0 & 1 & 1 \end{bmatrix} \oplus \begin{bmatrix} 0 & 1 & 1 \end{bmatrix} = \begin{bmatrix} 0 \oplus 0 & 1 \oplus 1 \end{bmatrix} = \begin{bmatrix} 0 & 0 & 0 \end{bmatrix}$  Fact:  $\underline{\ltimes} \oplus \underline{\ltimes} = \underline{O}$

g. 
$$\begin{bmatrix} 1 & 0 & 1 \\ 0 & 1 \\ 1 & 0 \end{bmatrix} = \begin{bmatrix} (1 \cdot 1) \oplus (0 \cdot 0) \oplus (1 \cdot 1) & (1 \cdot 0) \oplus (0 \cdot 1) \oplus (1 \cdot 0) \end{bmatrix} = \begin{bmatrix} 0 & 0 \end{bmatrix}$$
  
Alternatively,  $\begin{bmatrix} 1 & 0 & 1 \end{bmatrix} \begin{bmatrix} \mu^{(1)} \\ \mu^{(2)} \\ \mu^{(3)} \end{bmatrix} = \begin{bmatrix} 0 & 0 \end{bmatrix}$ 

# 2. A codeword $\begin{bmatrix} 1 & 0 & 1 \end{bmatrix}$ is sent over the BSC. Suppose the error pattern is $\underline{e} = \begin{bmatrix} 0 & 0 & 1 & 1 \end{bmatrix}$ . Find the observed vector at the receiver. $\chi = \underline{x} \oplus \underline{e} = \begin{bmatrix} 1 & 1 & 0 & 1 \end{bmatrix} \oplus \begin{bmatrix} 0 & 0 & 1 & 1 \end{bmatrix} = \begin{bmatrix} 1 & 1 & 1 & 0 \end{bmatrix}$ . means that the last two bits of $\underline{x}$ are received

4. A codeword is sent over the BSC. Suppose the observed vector at the receiver is  $\mathbf{y} = \begin{bmatrix} 0 & 1 & 1 & 1 \end{bmatrix}$  and the error pattern is  $\mathbf{e} = \begin{bmatrix} 0 & 0 & 1 & 1 \end{bmatrix}$ .  $\mathbf{z} \in \mathbf{z}$  are differnt.

$$\chi = \underbrace{\times} \bigoplus \underbrace{e} = \begin{bmatrix} 0 & 1 & 1 & 1 \end{bmatrix} \bigoplus \begin{bmatrix} 0 & 0 & 1 & 1 \\ 0 & 0 & 1 & 1 \end{bmatrix} \bigoplus \begin{bmatrix} 0 & 0 & 1 & 1 \end{bmatrix} \bigoplus \begin{bmatrix} 0 & 0 & 1 & 1 \end{bmatrix} \bigoplus \begin{bmatrix} 0 & 0 & 1 & 1 \end{bmatrix} \bigoplus \begin{bmatrix} 0 & 0 & 1 & 0 \end{bmatrix}$$

Alternatively, the error pattern says that the last two bits in the observed vector are received incorrectly. So, we need to flip their values to get the transmitted codeword.

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c.

Method :

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1. A linear block code uses the following generator matrix  $\,G=\,$ 



- a. Find the code length *n* The generator marix has 4 columns. **n** = 4
- b. Find the codeword for the message  $\mathbf{b} = \begin{bmatrix} 0 & 1 \end{bmatrix}$

b. 9

Find the codebook for this code.

**x** =

$$\mathbf{x} = \mathbf{b} \mathbf{G} = \begin{bmatrix} 0 & 1 \end{bmatrix} \begin{bmatrix} 0 & 1 & 1 & 1 \\ 1 & 0 & 0 & 1 \end{bmatrix} = \begin{bmatrix} 1 & 0 & 0 & 1 \end{bmatrix}$$

where

Direct multiplication like this is
 OK when we only have to find one codeword. However, in the next part, we need to find all of them; so we will use other
 methods.

- Each row of the table below correspond to a code that uses single-parity-check. The error pattern is given.

Find the corresponding values of codeword length n, code dimension k, and Indicate whether the given error pattern is detectable.

Error Pattern	codeword length <i>n</i>	code dimension k	<u>e</u> is detectable (Yes or No?)
$\mathbf{\underline{e}} = \begin{bmatrix} 1 & 0 & 0 & 1 \end{bmatrix}$	4	3	No
$\mathbf{\underline{e}} = \begin{bmatrix} 1 & 0 & 1 & 0 & 1 \end{bmatrix}$	5	4	Yes
$\mathbf{\underline{e}} = \begin{bmatrix} 0 & 0 & 0 & 0 & 1 \end{bmatrix}$	5	4	Yes
$\underline{\mathbf{e}} = \begin{bmatrix} 1 & 1 & 1 & 1 & 1 & 1 & 0 \end{bmatrix}$	7	6	No

The length of vectors  

$$\underline{e}, \underline{x}, \underline{Y}$$
  
should all be the same  
 $(\underline{Y} = \underline{x} \oplus \underline{e})$   
So,  $n = \text{length of } \underline{e}$ 

For single-parity-check code,

× 1s

6

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Consider a block code whose generator matrix is

$$\mathbf{G} = \begin{pmatrix} 1 & 0 & 0 & 1 & 0 & 1 \\ 0 & 1 & 0 & 0 & 1 & 1 \\ 0 & 0 & 1 & 1 & 1 & 0 \end{pmatrix}$$

1. Find the code length *n* 

The #columns of G is 6

2. Find the length *k* of each message block

#### The **#rows of G is 3**

In the table below, list all possible data (message) vectors <u>b</u> in the leftmost column (one in each row).
 Then, find the corresponding codewords <u>x</u> and their weights in the second and third columns, respectively.



4. Find the minimum distance  $d_{\min}$  for this code.

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- 5. Do not panic.
- 1. Consider a linear block code that uses parity checking on a square array:



Each parity bit  $p_i$  is calculated such that the corresponding row or column has even parity. Suppose the following bits arrangement is used in the codeword:

$$\mathbf{\underline{x}} = (b_1 \quad p_1 \quad p_2 \quad b_2 \quad b_3 \quad b_4 \quad p_3 \quad p_4).$$

Find the generator matrix G. a.

b. Find the codeword for the message 
$$\mathbf{b} = \begin{bmatrix} 0 & 1 & 1 & 0 \end{bmatrix}$$
.

$$\underline{z} = \underline{b} = \begin{bmatrix} 0 & 1 & 1 & 0 \end{bmatrix} \begin{bmatrix} 9^{(1)} \\ 9^{(1)} \\ 9^{(3)} \\ 9^{(4)} \end{bmatrix} = \underbrace{9^{(2)} + 9^{(3)}}_{0 & 1 & 0 & 0 & 1 & 0 & 0 & 1 \\ 0 & 1 & 0 & 1 & 0 & 1 & 0 \\ 0 & 1 & 1 & 1 & 0 & 1 & 1 \end{bmatrix} \xrightarrow{0}$$

Identity matrix in the data positions becomes identity Find the parity-check matrix  ${f H}$ . c. matrix in the parity positions

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$G = \begin{bmatrix} 1 & 1 & 1 & 0 & 0 \\ 0 & 0 & 1 & 1 & 0 \\ 0 & 1 & 0 & 0 & 1 \\ 0 & 0 & 0 & 0 & 0 \end{bmatrix}$	0 0 0 0 0 1 0 1 0 1 1 1	1100 1011 0000 001	1000 0000 1110 0101	= H
and the second			••	
	Bit values in the p	oarity posi	tions are trai	nsposed and

put in the data positions

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Consider a block code whose generator matrix is

# rix is $\mathbf{G} = \begin{bmatrix} 1 & 0 & 0 \\ 0 & 1 & 0 \\ 0 & 0 & 1 \end{bmatrix} \begin{bmatrix} 1 & 0 & 1 \\ 0 & 1 & 1 \\ 1 & 1 & 0 \end{bmatrix}$

a. Find the parity check matrix **H** of this code.



ii. Find the decoded codeword <u>x</u>
The syndrome <u>A</u> is the same as the last column of H.
Therefore, <u>ê</u> = [0 0 0 0 01] and
<u>ê</u> = <u>x</u> - <u>ê</u> = <u>x</u> <u>⊕</u> <u>ê</u> = [111000]
iii. Find the decoded message <u>b</u>.
<u>b</u> = [11]
From <u>G</u>, we have I<sub>3</sub>
in the front, so
the message <u>b</u> will be the first
three bits of the codeword <u>x</u>.

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Consider a convolution encoder represented by the following diagram



(a) Draw the corresponding state (transition) diagram



First, observe that this encoder uses a shift register with two FFs which is the same as the one discussed in lecture. Therefore, the arrows will be the same as what we had in the lecture.

Note, however, that the connections that produce the outputs are different from the encoder in lecture. Therefore, we simply need to find the outputs

		30	<u></u>	
<b>b</b>	5,	2	*"	) <u>*</u> (1)
0	0	0	0	0
1	0	0	1	1
0	0	1	0	1
1	0	1	1	0
0	1	0	1	1
1	1	0	0	0
0	1	1	1	0
1	1	1	0	1

- (b) Suppose the information bits (the message bits) are b = 11101.
  - Find the corresponding codeword  $\underline{\mathbf{x}}$
  - i. by using the direct method (filling out the table below) without the help of the state diagram from part (a).

Note that the final output is one row vector resulting from interleaving the upper and lower outputs.



#### $\underline{x} = 1100011010$

#### and

ii. by "tracing" the corresponding path on the state diagram derived an part (a)Draw/highlight your trace on the state diagram in part (a) using different pen color.

See the trace in the diagram on the

 $\underline{\mathbf{x}} = \underline{1} \underline{1} \underline{0} \underline{0} \underline{0} \underline{1} \underline{1} \underline{0} \underline{1} \underline{0}$ 

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Consider a convolutional encoder whose trellis diagram is given below



1. Suppose the data vector is  $\underline{\mathbf{b}}$  = [110]. Find the corresponding codeword  $\underline{\mathbf{x}}$ .

## Read from the highlighted path: [110101]

2. Suppose that we observe  $\underline{\mathbf{y}}$  = 110110 at the input of the minimum distance decoder.

The decoder uses Viterbi's algorithm.

a. <u>Write down</u>

(1) all the distance values on the branches and

(2) the (chosen) cumulative distance values inside all the circles

in the figure above.

b. Find the decoded codeword  $\hat{\underline{x}}$  and the decoded message  $\hat{\underline{b}}$ .

Read from the highlighted path:  $\hat{x} = [110110]$   $\hat{b} = [111]$ .  $\hat{b}$ No error in the received vector

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#### Instructions

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- Do not panic. 3.

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1. Suppose 
$$\overline{v} = \begin{pmatrix} 4 \\ 0 \\ -3 \end{pmatrix}$$
 and  $\overline{u} = \begin{pmatrix} 2 \\ 1 \\ -2 \end{pmatrix}$ .  
a. Find  $\langle \overline{v}, \overline{u} \rangle = (4)(2) + (0)(1) + (-3)(-2) = 8+0+6 = 1$   
For  $\overline{v} = \langle v_1, v_2, v_3 \rangle^T$  and  $\widehat{u} = [u_1, u_2, u_3 \rangle^T$ ,  
we have  
 $\langle \overline{v}, \overline{u} \rangle = \sqrt{u_1} + \sqrt{u_2}u_2 + \sqrt{u_3}u_3$   
b. Find  $\langle \overline{u}, \overline{u} \rangle = 2^2 + 1^2 + (-2)^2 = 4 + 1 + 4 = 9$   
For  $\overline{u} = (u_2, u_2, u_3)^T$ ,  
we have  
 $\langle \overline{u}, \overline{u} \rangle = u_1u_1 + u_2u_2 + u_3u_3 = u_1^2 + u_2^2 + u_3^2$   
c. Find  $\langle \overline{v}, \overline{v} \rangle = \sqrt{4} + 0^2 + (-3)^2 = 16+9 = 25$   
Similar to part (b),  
for  $\overline{v} = (\sqrt{1}, v_2, \sqrt{3})^T$ ,  
we have  
 $\langle \overline{u}, \overline{v} \rangle = \sqrt{4} + \sqrt{2} + \sqrt{3}$   
d. Find  $\|\overline{v}\| = \sqrt{\langle \overline{v}, \overline{v} \rangle} = \sqrt{2} = 5$   
by definition

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e. Find  $\|\mathbf{\bar{u}}\| \leq \sqrt{2\omega_j}$ Find  $\operatorname{proj}_{\overline{u}} \overline{v} \equiv \langle \overline{v}, \overline{u} \rangle = \frac{14}{9} \begin{pmatrix} 2 \\ 1 \\ -2 \end{pmatrix} = \begin{pmatrix} 28/9 \\ 14/9 \\ -28/9 \end{pmatrix} \approx \begin{pmatrix} 3.1111 \\ 1.5556 \\ -3.1111 \end{pmatrix}$  $\begin{array}{c} \text{Direct calculation:} \\ \vec{z} = (4, 0, -3)^{T} - \frac{1}{4} (28, 14, -28)^{T} = \frac{1}{9} (8, -14, 1)^{T} \\ \text{g. Let } \vec{z} = \vec{v} - \text{proj}_{\vec{u}} \ \vec{v} \ \text{Find } \langle \vec{u}, \vec{z} \rangle \ (\vec{v}, \vec{z}) = \frac{1}{9} ((2)(1) + (1)(-14) + (-2)(1)) = \frac{1}{9} \times 0 = 0 \end{array}$ We know that  $\vec{p} = proj_{\vec{w}}$  and  $\vec{\sigma} = \vec{v} - \vec{p}$  are always orthogonal.  $\vec{x}$  is parallel to  $\vec{p}$  the same as  $\vec{\sigma}$ Therefore, it and i are orthogonal and hence < i, i > = 0.

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f.

$$\langle \vec{u}, \vec{z} \rangle = \langle \vec{u}, \vec{v} - \rho v_j \vec{u} \vec{v} \rangle = \langle \vec{u}, \vec{v} \rangle - \langle \vec{u}, \langle \vec{v}, \vec{u} \rangle = \langle \vec{u}, \vec{v} \rangle - \langle \vec{v}, \vec{u} \rangle = 0$$

$$\langle \vec{u}, \vec{u} \rangle = \langle \vec{u}, \vec{v} \rangle - \langle \vec{v}, \vec{u} \rangle = 0$$

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Consider the two signals  $s_1(t)$  and  $s_2(t)$  shown below.

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a. Find the energy of each signal.

$$E_{1} = \int_{-\infty}^{\infty} \beta_{1}^{2} (t) dt = \int_{2}^{2} 2^{2} dt = \int_{4}^{2} 4 dt - 4t \Big|_{0}^{2} = 4 \times 2 = 8$$
  

$$E_{2} = \int_{-\infty}^{\infty} \beta_{2}^{2} (t) dt = \int_{0}^{4} (\frac{t}{2})^{2} dt = \int_{0}^{4} \frac{t^{2}}{4} dt = \frac{t^{3}}{12} \Big|_{0}^{4} = \frac{4 \times 4 \times 4}{12} = \frac{16}{3}$$

b. Find their inner product  $\langle s_1(t), s_2(t) \rangle$ .

$$\langle \mathfrak{d}_{1}(t), \mathfrak{d}_{2}(t) \rangle = \int_{-\infty}^{\infty} \mathfrak{d}_{1}(t) \mathfrak{d}_{2}(t) dt = \int_{0}^{\infty} \mathfrak{d}_{1}(t) \mathfrak{d}_{2}(t) dt = \int_{0}^{\infty} \mathfrak{d}_{1}(t) \mathfrak{d}_{2}(t) dt = \int_{0}^{\infty} \mathfrak{d}_{1}(t) \mathfrak{d}_{2}(t) dt = \frac{t^{2}}{2} = \frac{t^{2}}$$

c. Find and plot  $\operatorname{proj}_{s_1(t)} s_2(t)$ .



## ECS 452: In-Class

Date:

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#### Continue from Exercise #18.

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d. Suppose the Gram-Schmidt orthogonalization procedure (GSOP) (where the signals are applied in **the order given**) is used to find two orthonormal functions  $\phi_1(t)$  and  $\phi_2(t)$  that can be used as axes to represent  $s_1(t)$  and  $s_2(t)$ . i. Find and plot  $\phi_1(t)$ Ø,LE>  $w_1(t) = \delta_1(t)$   $\delta_1 = 2\sqrt{2} \beta_1 \approx 2.83 \beta_1$  $\mathcal{P}_{1}(t) = \frac{\omega_{1}(t)}{\sqrt{E_{\omega}}} = \frac{\mathcal{D}_{1}(t)}{\sqrt{E_{\omega}}} = \frac{1}{2\sqrt{E_{\omega}}} \mathcal{D}_{1} = \begin{cases} 1/\sqrt{2}, & 0.4 \le 2\\ 0, & 0.4 \end{cases}$ ii. Find and plot  $\phi_2(t)$ .  $u_2(t) = \partial_1(t) - proj_{u_1} \partial_2 = \partial_2(t) - \frac{1}{4} \partial_1(t)$  $E_{u_{2}} = \int u_{2}^{2} (t) dt = \int \left(-\frac{1}{2} + \frac{t}{2}\right)^{2} dt + \int \left(\frac{t}{2}\right)^{2} dt$  $\varphi_2(t)$  $=\frac{1}{6}+\frac{14}{3}=\frac{1+28}{6}=\frac{29}{6}$   $\int_{2}^{3} (t-1) \quad 0 < t < 2,$   $\int_{2}^{3} (t-1) \quad 0 < t < 2,$   $\int_{29}^{3} (t-1) \quad 0 < t < 2,$   $\int_{38}^{3} (t-1) \quad 0 < t < 2,$   $\int_{58}^{3} (t-1) \quad 0 < t$  $\mathcal{Q}(t) = \frac{1}{4} \mathcal{A}_{1}(t) + \mathcal{U}_{2}(b) = \frac{1}{4} 2 \sqrt{2} \mathcal{A}_{1}(t) + \sqrt{\frac{29}{6}} \mathcal{A}_{2}(t) = \frac{1}{\sqrt{2}} \mathcal{A}_{1}(t) + \sqrt{\frac{29}{6}} \mathcal{A}_{2}(t) \approx 0.71 \mathcal{A}_{1}(t) + 2.2 \mathcal{A}_{2}(t)$ iii. Find the two vectors,  $\vec{s}^{(1)}$  and  $\vec{s}^{(2)}$ , that represent the two waveforms  $s_1(t)$  and  $s_2(t)$  in the new axes based on  $\phi_1(t)$  and  $\phi_2(t)$ . Draw the corresponding constellation in the figure below below.



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A digital communication system transmits a stream of bits by mapping each block of three bits to one of the possible waveforms  $s_1(t), s_2(t), ..., s_M(t)$ . The waveform is then transmitted via a communication channel which corrupts the waveform by independently adding a white noise process N(t) whose power spectral density is given by  $S_N(f) = 16$  across all frequency.

- a. What is the value of *M*?
  - Three bits ⇒ 8 possibilities for the block

⇒ need 8 different waveforms to represent 8 distinct block patterns

b. Suppose we apply GSOP to the *M* waveforms and get two orthonormal axes  $\phi_1(t)$  and  $\phi_2(t)$ . Let  $N_j = \langle N(t), \phi_j(t) \rangle$ . Find i.  $\mathbb{E}[N_1] = \mathbf{O}$ 

See 7.26f

- ii.  $\operatorname{Var}[N_1] = \frac{N_0}{2} = 16$
- iii.  $\sigma_{N_1} = \sqrt{Var[N_1]} = \sqrt{16} = 4$
- iv.  $\mathbb{E}[N_1N_2]$

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In a binary antipodal signaling scheme, the message S is randomly selected from the alphabet set  $S = \{-2, 2\}$  with  $p_1 = P[S = -2] = 0.6$  and  $p_2 = P[S = 2] = 0.4$ . The message is corrupted by an independent additive noise N whose pdf is shown below:



- a. What is the value of *h*?
  - To be a pdf, we need  $\int_{-\infty}^{\infty} f_{N}(n) dn = 1.$  $\frac{1}{2} \times 8 \times h = 1 \Rightarrow h = \frac{1}{4}$
- b. Suppose the received symbol is R = r. Find the MAP detector  $\hat{s}_{MAP}(r)$ .

