

## 6.2 Signal Space Concepts

As in the case of vectors, we now develop a parallel treatment for a set of signals.

### Definition 6.36.

- (a) The **inner product** of two generally complex-valued signals  $x_1(t)$  and  $x_2(t)$  is denoted by  $\langle x_1(t), x_2(t) \rangle$  and defined by

$$\langle x_1, x_2 \rangle = \langle x_1(t), x_2(t) \rangle = \int_{-\infty}^{\infty} x_1(t)x_2^*(t)dt.$$

- (b) The signals are **orthogonal** if their inner product is zero.

- (c) The **norm** of a signal is defined as

$$\|x\| = \|x(t)\| = \sqrt{\langle x(t), x(t) \rangle} = \sqrt{E_x}$$

where  $E_x$  is the energy in  $x(t)$ :

$$E_x = \langle x(t), x(t) \rangle = \int_{-\infty}^{\infty} x(t)x^*(t) dt = \int_{-\infty}^{\infty} |x(t)|^2 dt$$

- (d) A collection of  $N$  signals is **orthonormal** if the signals are orthogonal and their norms are all unity.

**Example 6.37.** Consider the two waveforms shown in Figure 15.

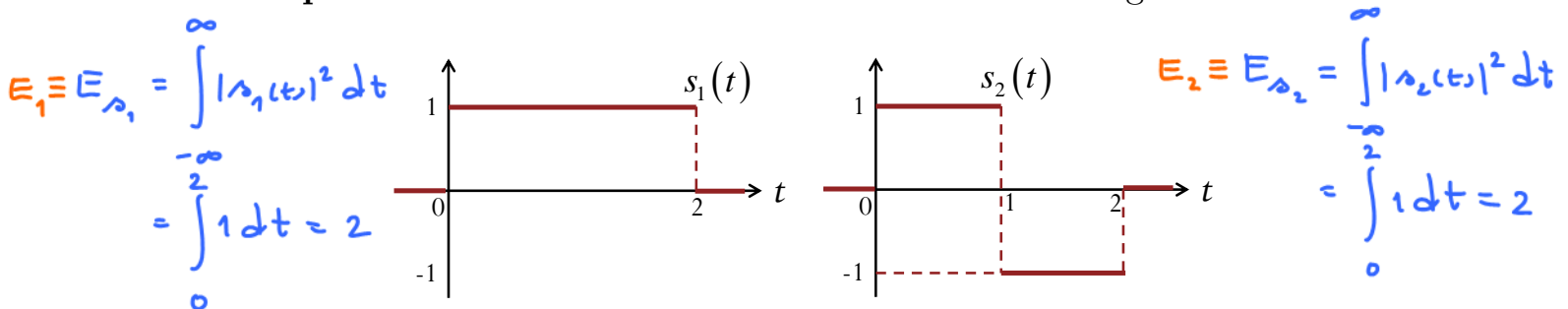
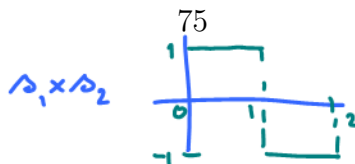


Figure 15: Two Waveforms in Example 6.37

$$\|s_1\| = \sqrt{E_1} = \sqrt{2}$$

$$\|s_2\| = \sqrt{E_2} = \sqrt{2}$$

$$\langle s_1(t), s_2(t) \rangle = \int_{-\infty}^{\infty} s_1(t)s_2^*(t)dt = 0$$



**Definition 6.38.**

(a) The **projection** of  $x_2(t)$  to  $x_1(t)$  is given by

$$\text{proj}_{x_1(t)} x_2(t) = \frac{\langle x_2, x_1 \rangle}{\langle x_1, x_1 \rangle} x_1 = \frac{\langle x_2(t), x_1(t) \rangle}{\langle x_1(t), x_1(t) \rangle} x_1(t) = \frac{\langle x_2(t), x_1(t) \rangle}{E_{x_1}} x_1(t)$$

(b) The **cross-correlation coefficient** of  $x_1(t)$  and  $x_2(t)$  is defined as

$$\rho_{x_1, x_2} = \frac{\langle x_1(t), x_2(t) \rangle}{\sqrt{E_{x_1} E_{x_2}}}$$

- $\text{proj}_{x_1(t)} x_2(t) = \sqrt{E_{x_2}} \rho_{x_2, x_1} \frac{x_1(t)}{\sqrt{E_{x_1}}}$

**Example 6.39.** For the two waveforms shown in Figure 15,

$$\text{proj}_{s_1} s_2 = \frac{0}{\infty} s_1 = 0 s_1 = 0$$

← this is not just a number. This is a function/waveform that is 0 all the time.

**6.40.** Similar to 6.31, the **Gram-Schmidt Orthogonalization Procedure (GSOP)** can be used to construct a set of orthonormal waveforms from a set of finite energy signal waveforms:  $\{s_j(t), j = 1, 2, \dots, M\}$ .

The first orthonormal function is simply constructed as

$$\phi_1(t) = \frac{u_1(t)}{\sqrt{E_{u_1}}} = \frac{s_1(t)}{\sqrt{E_{s_1}}}$$

The subsequent orthonormal functions are found as follows:

$$\phi_i(t) = \frac{u_i(t)}{\sqrt{E_{u_i}}}$$

where the unnormalized basis function  $u_i(t)$  is given by

$$u_i(t) = s_i(t) - \sum_{k=1}^{i-1} \text{proj}_{u_k(t)} s_i(t)$$

and

$$\text{proj}_{u_k(t)} s_i(t) = \frac{\langle s_i(t), u_k(t) \rangle}{\langle u_k(t), u_k(t) \rangle} u_k(t) = \langle s_i(t), \phi_k(t) \rangle \phi_k(t)$$

As with the GSOP for vectors, we also discard the zero functions. In general, the final number of orthonormal functions,  $N$ , is less than or equal to the number of given waveforms,  $M$ , depending on one of the two possibilities:

- (a) If the waveforms  $\{s_j(t), j = 1, 2, \dots, M\}$  form a linearly independent set, then  $N = M$ .
- (b) If the waveforms  $\{s_j(t), j = 1, 2, \dots, M\}$  are not linearly independent, then  $N < M$ .

**Example 6.41.** Consider the four waveforms illustrated in Figure 16. Use the Gram-Schmidt orthogonalization procedure (where the waveforms are applied in the order given) to find the orthonormal basis waveforms  $\phi_1(t), \phi_2(t), \dots$  whose linear combinations can be used to represent the four waveforms.

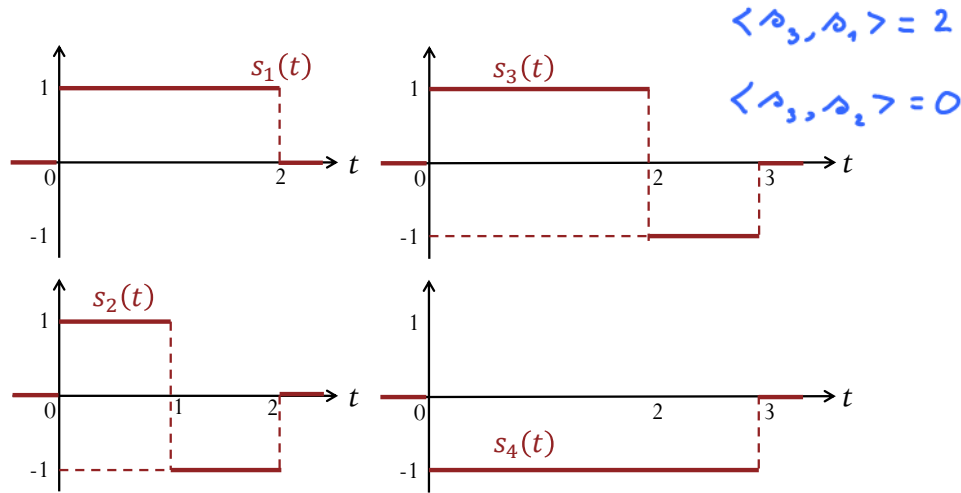


Figure 16: Four signals for orthogonalization in Example 6.41

Handwritten calculations for the Gram-Schmidt process:

$u_1 = \rho_1$

$E_{u_1} = E_{\rho_1} = 2$  (Ex. 6.37)

$\phi_1(t) = \frac{u_1}{\sqrt{E_{u_1}}} = \frac{1}{\sqrt{2}} u_1(t) = \frac{1}{\sqrt{2}} \rho_1(t)$

$u_2 = \rho_2 - \text{proj}_{u_1} \rho_2 = \rho_2 - \frac{\langle \rho_2, u_1 \rangle}{\langle u_1, u_1 \rangle} u_1 = \rho_2 - \frac{\langle \rho_2, \rho_1 \rangle}{\langle \rho_1, \rho_1 \rangle} \rho_1 = \rho_2 - \frac{0}{2} \rho_1 = \rho_2$  (Ex. 6.39)

$E_{u_2} = E_{\rho_2} = 2$

$\phi_2(t) = \frac{u_2}{\sqrt{E_{u_2}}} = \frac{1}{\sqrt{2}} u_2 = \frac{1}{\sqrt{2}} \rho_2(t)$

$\langle \rho_3, u_2 \rangle = \langle \rho_3, \rho_2 \rangle = 0$

$u_3 = \rho_3 - \text{proj}_{u_1} \rho_3 - \text{proj}_{u_2} \rho_3 = \rho_3 - \frac{\langle \rho_3, u_1 \rangle}{\langle u_1, u_1 \rangle} u_1 - \frac{\langle \rho_3, u_2 \rangle}{\langle u_2, u_2 \rangle} u_2 = \rho_3 - \frac{\langle \rho_3, \rho_1 \rangle}{\langle \rho_1, \rho_1 \rangle} \rho_1 = \rho_3 - \frac{2}{2} \rho_1 = \rho_3 - \rho_1$  (Ex. 6.39)

$E_{u_3} = 1$

$\phi_3(t) = \frac{u_3}{\sqrt{E_{u_3}}} = u_3$

$u_3(t) = \phi_3(t)$

77

~~$s_4 = s_4 - \text{proj}_{m_1} s_4 - \text{proj}_{m_2} s_4 - \text{proj}_{m_3} s_4 = 0$~~   
 discarded

**6.42.** Once we have constructed<sup>16</sup> the set of, say  $N$ , orthonormal waveforms  $\{\phi_i(t), i = 1, 2, \dots, N\}$ , we can express the signals  $s_i(t)$  as linear combinations of the  $N$  orthonormal basis functions  $\phi_i(t)$ . Thus, we may write

$$s_j(t) = \sum_{i=1}^N s_i^{(j)} \phi_i(t) \quad (33)$$

$s_j(t) = \underbrace{[\phi_1(t) \ \phi_2(t) \ \dots \ \phi_N(t)]}_{\Phi(t)} \begin{pmatrix} s_1^{(j)} \\ s_2^{(j)} \\ \vdots \\ s_N^{(j)} \end{pmatrix}$

where the constants (weights)

$$s_i^{(j)} = \langle s_j(t), \phi_i(t) \rangle. \quad (34)$$

Note that  $s_i^{(j)} \phi_i(t) = \langle s_j(t), \phi_i(t) \rangle \phi_i(t)$  can be geometrically interpreted as the projection of the signal  $s_j(t)$  onto the  $i$ th axis,  $\phi_i(t)$ .

Based on (33), each signal may be represented by the vector (or sequence)

$$\mathbf{s}^{(j)} = (s_1^{(j)}, s_2^{(j)}, \dots, s_N^{(j)})^T, \quad (35)$$

or, equivalently, as a point in the  $N$ -dimensional (in general, complex) signal space.

The (mathematical/conceptual) conversion/mapping from waveform to its corresponding vector in (35) and (34) is shown in Figure 17a. The inverse mapping from vector to waveform in (33) is shown in Figure 17b.

**Example 6.43.** For the four waveforms in Example 6.41 and the orthonormal basis derived from GSOP,

$$\begin{aligned}
 s_1(t) &= \sqrt{2} \phi_1(t) &= \Phi(t) \begin{pmatrix} \sqrt{2} \\ 0 \\ 0 \end{pmatrix} &\rightarrow \mathbf{s}^{(1)} \\
 s_2(t) &= \sqrt{2} \phi_2(t) &= \Phi(t) \begin{pmatrix} 0 \\ \sqrt{2} \\ 0 \end{pmatrix} &\rightarrow \mathbf{s}^{(2)} \\
 s_3(t) &= \sqrt{2} \phi_1(t) + \phi_3(t) &= \Phi(t) \begin{pmatrix} \sqrt{2} \\ 0 \\ 1 \end{pmatrix} &\rightarrow \mathbf{s}^{(3)} \\
 s_4(t) &= -\sqrt{2} \phi_1(t) + \phi_3(t) &= \Phi(t) \begin{pmatrix} -\sqrt{2} \\ 0 \\ 1 \end{pmatrix} &\rightarrow \mathbf{s}^{(4)}
 \end{aligned}$$

<sup>16</sup>We have shown how this set can be constructed from GSOP. However, in practice, this set may be derived from different procedure.

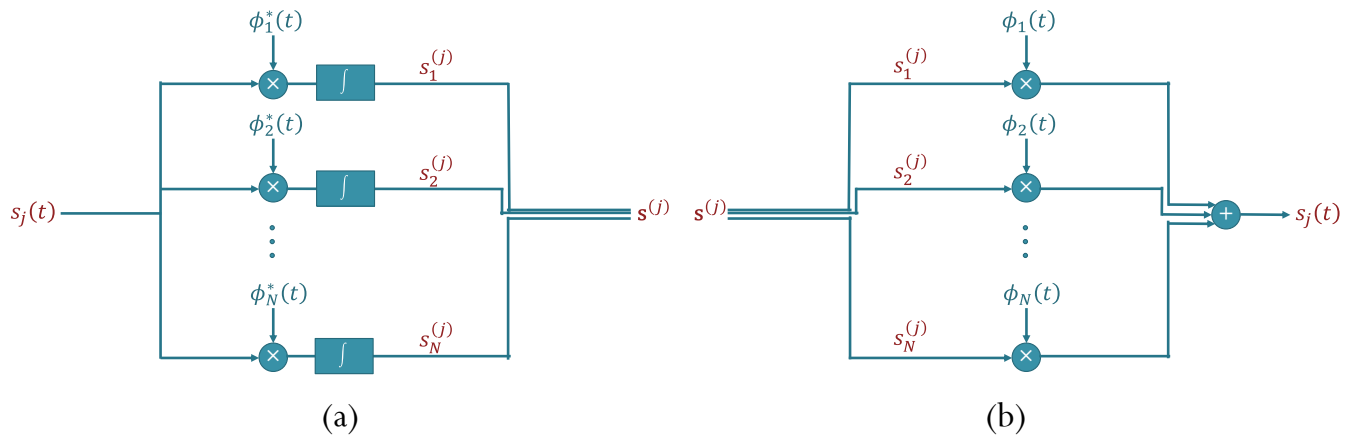


Figure 17: Waveform to vector (a), and vector to waveform (b) mappings.

**Definition 6.44.** From 6.42, a set of  $M$  signals  $\{s_j(t), j = 1, 2, \dots, M\}$  can be represented by a set of  $M$  vectors  $\{\mathbf{s}^{(j)}\}$  in the  $N$ -dimensional space. The corresponding set of vectors is called the **signal space representation**, or **constellation**, of  $\{s_j(t), j = 1, 2, \dots, M\}$ .

**6.45.** From the orthonormality of the basis, we have

- (a) the inner product of two signals is equal to the inner product of the corresponding vectors:

$$\langle s_i(t), s_j(t) \rangle = \langle \mathbf{s}^{(i)}, \mathbf{s}^{(j)} \rangle.$$

$$\langle s_1(t), s_2(t) \rangle = \langle \vec{s}^{(1)}, \vec{s}^{(2)} \rangle = 0$$

(b)  $E_j \equiv E_{s^{(j)}} = \|s_j(t)\|^2 = \|\mathbf{s}^{(j)}\|^2.$

$$E_{s_1} = \langle s_1(t), s_1(t) \rangle = \int_{-\infty}^{\infty} |s_1(t)|^2 dt = \langle \vec{s}^{(1)}, \vec{s}^{(1)} \rangle = 2 + 0 + 0 = 2$$

**6.46.** It should be emphasized, however, that the functions  $\{\phi_i(t)\}$  obtained from the Gram-Schmidt procedure are not unique. If we alter the order in which the orthogonalization of the signals  $\{s_j(t)\}$  is performed, the orthonormal waveforms will be different and the corresponding vector representation of the signals  $\{s_j(t)\}$  will depend on the choice of the orthonormal functions  $\{\phi_i(t)\}$ . Nevertheless, the dimensionality of the signal space ( $N$ ) will not change, and the vectors  $\mathbf{s}^{(j)}$  will **retain** their **geometric configuration**; i.e., their lengths and their inner products will be invariant to the choice of the orthonormal functions  $\{\phi_i(t)\}$ .