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## ECS 452: Digital Communication Systems

2019/2

HW 3 — Due: Feb 18, 4 PM

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## Instructions

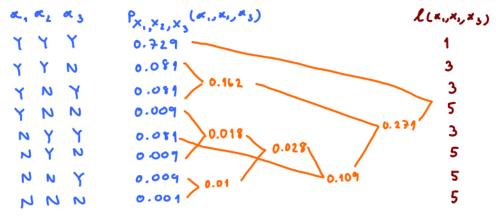
(a) This assignment has 6 pages.

- (b) (1 pt) Hard-copies are distributed in class. Original pdf file can be downloaded from the course website. Work and write your answers <u>directly on the provided hardcopy/file</u> (not on other blank sheet(s) of paper).
- (c) (1 pt) Write your first name and the last three digits of your student ID in the spaces provided on the upper-right corner of this page. Furthermore, for online submission, your file name should start with your 10-digit student ID, followed by a space, the course code, a space, and the assignment number: "5565242231 452 HW3.pdf"
- (d) (8 pt) Try to solve all non-optional problems.
- (e) Late submission will be heavily penalized.
- (f) Write down all the steps that you have done to obtain your answers. You may not get full credit even when your answer is correct without showing how you get your answer.

**Problem 1.** Continue from the Example 2.40 in the lecture. A memoryless source emits two possible message Y(es) and N(o) with probability 0.9 and 0.1, respectively.

(a) Calculate the entropy (per source symbol) of this source.

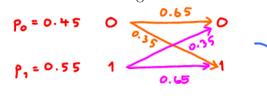
(b) Find the expected codeword length per symbol of the Huffman binary code for the third-order extensions of this source.



E[L(x,x,x3)] = 1.5980 bits per three source symbols => L3 = 0.5327 bits per source symbols

**Problem 2.** Consider a BSC whose crossover probability for each bit is p = 0.35. Suppose 1-P = 0.65 P[X=0] = 0.45.

 $p_1 = 1 - p_0 = 1 - 0.45 = 0.55$ (a) Draw the channel diagram.



 $\hat{x}(y) = \hat{x}(0) = \hat{x}(1)$ 

(b) Find the channel matrix **Q**. Method 1 ·

(c) Find the joint pmf matrix **P**.

To get the matrix P, we simply scale each row of matrix Q by the corresponding p(x)

the matrix P, we simply each row of matrix Q by responding p(x)

(d) Find the row vector 
$$\underline{\mathbf{q}}$$
 which contains the pmf of the channel output Y.

Method 1: 
$$_{9} = [0.485 \ 0.515]$$

 $P(\mathcal{E}) = A - P(\mathcal{C})$ 

Method 2: Directly read the transition (conditional)

Method 2: 
$$Q = PQ = [0.45 \ 0.55] \begin{bmatrix} 0.65 \ 0.35 \\ 0.35 \ 0.65 \end{bmatrix} = [0.485 \ 0.515]$$

(e) Analyze the performance of all four reasonable detectors for this binary channel. Complete the table below:

 $\overline{P(C)}$ 

		x(y)	x(0)	x(1)	I(C)	I (C) -	15 5 (4)
		y	0	1	0.65(Po+P1) = 0.65	5 0.35	
r	٦.	1-y	1	0	0.35(Po+P1) = 0.3	5 6. 45	
0 - 11-P	P	1	1	1	(0.35+0.65)P1=P1=0	0.55	
Q = 1-p		0	0	0	(0.65+0-35) Po= Po=	0.45 0.55	
L P	1-P]			~_			
\Y 0	4	S 1/			The v	ralue of sely)	tell us which rom
N/	- '\c7 ×	P. ">	0	, _1			
0 0.6	5 0.35 -	$\rightarrow$	.65p.	0,3	[Po] - D to s	elect in the	column y of the P
2	1 5 0.35 — 55 0.65 — ×				mat		
1 [0-3	55 0.65]-	→ 1[6	.35 P	) ( <u>0.                                   </u>	5 P()		
	×	<u>ا</u>			2 2		

**Problem 3.** Consider a BAC whose Q(1|0) = 0.35 and Q(0|1) = 0.55. Suppose P[X = 0] =⇒ P1= 1-0.4 =0.6 0.4. **- P**•

In some of these parts, we will also try to derive the answer in a general form.

(a) Draw the channel diagram. So, we will start with Q(110) = a and Q(011) =/3.



To get the matrix P, we simply scale each row of matrix Q by the

(b) Find the joint pmf matrix P. we simply scale each recorresponding p(x).

First, we find the Q matrix,  $Q = \begin{pmatrix} 1 - \alpha & \alpha \\ 1 - \beta & 1 - \beta \end{pmatrix} = \begin{pmatrix} 1 - \alpha & \alpha \\ 1 - \beta & 1 - \beta \end{pmatrix} = \begin{pmatrix} 1 - \alpha & \alpha \\ 1 - \beta & 1 - \beta \end{pmatrix} = \begin{pmatrix} 1 - \alpha & \alpha \\ 1 - \beta & 1 - \beta \end{pmatrix} = \begin{pmatrix} 1 - \alpha & \alpha \\ 1 - \beta & 1 - \beta \end{pmatrix} = \begin{pmatrix} 1 - \alpha & \alpha \\ 1 - \beta & 1 - \beta \end{pmatrix} = \begin{pmatrix} 1 - \alpha & \alpha \\ 1 - \beta & 1 - \beta \end{pmatrix} = \begin{pmatrix} 1 - \alpha & \alpha \\ 1 - \beta & 1 - \beta \end{pmatrix} = \begin{pmatrix} 1 - \alpha & \alpha \\ 1 - \beta & 1 - \beta \end{pmatrix} = \begin{pmatrix} 1 - \alpha & \alpha \\ 1 - \beta & 1 - \beta \end{pmatrix} = \begin{pmatrix} 1 - \alpha & \alpha \\ 1 - \beta & 1 - \beta \end{pmatrix} = \begin{pmatrix} 1 - \alpha & \alpha \\ 1 - \beta & 1 - \beta \end{pmatrix} = \begin{pmatrix} 1 - \alpha & \alpha \\ 1 - \beta & 1 - \beta \end{pmatrix} = \begin{pmatrix} 1 - \alpha & \alpha \\ 1 - \beta & 1 - \beta \end{pmatrix} = \begin{pmatrix} 1 - \alpha & \alpha \\ 1 - \beta & 1 - \beta \end{pmatrix} = \begin{pmatrix} 1 - \alpha & \alpha \\ 1 - \beta & 1 - \beta \end{pmatrix} = \begin{pmatrix} 1 - \alpha & \alpha \\ 1 - \beta & 1 - \beta \end{pmatrix} = \begin{pmatrix} 1 - \alpha & \alpha \\ 1 - \beta & 1 - \beta \end{pmatrix} = \begin{pmatrix} 1 - \alpha & \alpha \\ 1 - \beta & 1 - \beta \end{pmatrix} = \begin{pmatrix} 1 - \alpha & \alpha \\ 1 - \beta & 1 - \beta \end{pmatrix} = \begin{pmatrix} 1 - \alpha & \alpha \\ 1 - \beta & 1 - \beta \end{pmatrix} = \begin{pmatrix} 1 - \alpha & \alpha \\ 1 - \beta & 1 - \beta \end{pmatrix} = \begin{pmatrix} 1 - \alpha & \alpha \\ 1 - \beta & 1 - \beta \end{pmatrix} = \begin{pmatrix} 1 - \alpha & \alpha \\ 1 - \beta & 1 - \beta \end{pmatrix} = \begin{pmatrix} 1 - \alpha & \alpha \\ 1 - \beta & 1 - \beta \end{pmatrix} = \begin{pmatrix} 1 - \alpha & \alpha \\ 1 - \beta & 1 - \beta \end{pmatrix} = \begin{pmatrix} 1 - \alpha & \alpha \\ 1 - \beta & 1 - \beta \end{pmatrix} = \begin{pmatrix} 1 - \alpha & \alpha \\ 1 - \beta & 1 - \beta \end{pmatrix} = \begin{pmatrix} 1 - \alpha & \alpha \\ 1 - \beta & 1 - \beta \end{pmatrix} = \begin{pmatrix} 1 - \alpha & \alpha \\ 1 - \beta & 1 - \beta \end{pmatrix} = \begin{pmatrix} 1 - \alpha & \alpha \\ 1 - \beta & 1 - \beta \end{pmatrix} = \begin{pmatrix} 1 - \alpha & \alpha \\ 1 - \beta & 1 - \beta \end{pmatrix} = \begin{pmatrix} 1 - \alpha & \alpha \\ 1 - \beta & 1 - \beta \end{pmatrix} = \begin{pmatrix} 1 - \alpha & \alpha \\ 1 - \beta & 1 - \beta \end{pmatrix} = \begin{pmatrix} 1 - \alpha & \alpha \\ 1 - \beta & 1 - \beta \end{pmatrix} = \begin{pmatrix} 1 - \alpha & \alpha \\ 1 - \beta & 1 - \beta \end{pmatrix} = \begin{pmatrix} 1 - \alpha & \alpha \\ 1 - \beta & 1 - \beta \end{pmatrix} = \begin{pmatrix} 1 - \alpha & \alpha \\ 1 - \beta & 1 - \beta \end{pmatrix} = \begin{pmatrix} 1 - \alpha & \alpha \\ 1 - \beta & 1 - \beta \end{pmatrix} = \begin{pmatrix} 1 - \alpha & \alpha \\ 1 - \beta & 1 - \beta \end{pmatrix} = \begin{pmatrix} 1 - \alpha & \alpha \\ 1 - \beta & 1 - \beta \end{pmatrix} = \begin{pmatrix} 1 - \alpha & \alpha \\ 1 - \beta & 1 - \beta \end{pmatrix} = \begin{pmatrix} 1 - \alpha & \alpha \\ 1 - \beta & 1 - \beta \end{pmatrix} = \begin{pmatrix} 1 - \alpha & \alpha \\ 1 - \beta & 1 - \beta \end{pmatrix} = \begin{pmatrix} 1 - \alpha & \alpha \\ 1 - \beta & 1 - \beta \end{pmatrix} = \begin{pmatrix} 1 - \alpha & \alpha \\ 1 - \beta & 1 - \beta \end{pmatrix} = \begin{pmatrix} 1 - \alpha & \alpha \\ 1 - \beta & 1 - \beta \end{pmatrix} = \begin{pmatrix} 1 - \alpha & \alpha \\ 1 - \beta & 1 - \beta \end{pmatrix} = \begin{pmatrix} 1 - \alpha & \alpha \\ 1 - \beta & 1 - \beta \end{pmatrix} = \begin{pmatrix} 1 - \alpha & \alpha \\ 1 - \beta & 1 - \beta \end{pmatrix} = \begin{pmatrix} 1 - \alpha & \alpha \\ 1 - \beta & 1 - \beta \end{pmatrix} = \begin{pmatrix} 1 - \alpha & \alpha \\ 1 - \beta & 1 - \beta \end{pmatrix} = \begin{pmatrix} 1 - \alpha & \alpha \\ 1 - \beta & 1 - \beta \end{pmatrix} = \begin{pmatrix} 1 - \alpha & \alpha \\ 1 - \beta & 1 - \beta \end{pmatrix} = \begin{pmatrix} 1 - \alpha & \alpha \\ 1 - \beta & 1 - \beta \end{pmatrix} = \begin{pmatrix} 1 - \alpha & \alpha \\ 1 - \beta & 1 - \beta \end{pmatrix} = \begin{pmatrix} 1 - \alpha & \alpha \\ 1 - \beta & 1 - \beta \end{pmatrix} = \begin{pmatrix} 1 - \alpha & \alpha \\ 1 - \beta & 1 - \beta \end{pmatrix} = \begin{pmatrix} 1 - \alpha & \alpha \\ 1 - \beta & 1 - \beta \end{pmatrix} = \begin{pmatrix} 1 - \alpha & \alpha \\ 1 - \beta & 1 - \beta \end{pmatrix} = \begin{pmatrix} 1 - \alpha & \alpha \\$ 

(c) Find the row vector  $\underline{\mathbf{q}}$  which contains the pmf of the changel output Y.

Method 1: Recall that g = pQ.

For this question, 
$$p = [p \cdot 1 - p] = [0.4 \cdot 0.6]$$
.  
Therefore,  $q_{g} = [0.4 \cdot 0.6] \begin{bmatrix} 0.26 & 0.14 \\ 0.32 & 0.27 \end{bmatrix} = \begin{bmatrix} 0.59 & 0.41 \end{bmatrix}$ 

(d) Analyze the performance of all four reasonable detectors for this binary channel. Complete the table below:

$\hat{x}(y)$	$\hat{x}(0)$	$\hat{x}(1)$	P(C)	$P(\mathcal{E})$ =	1-P(C)
y	0	1	0.26+0.27=0.53	0.47	
1-y	1	0	0.33+0.14 = 0.47	0.53	
1	1	1	0.33+0.27 = 0.60	0.40	
0	0	0	0.26+0.14 = 0.40	0.60	

$$Q = \begin{bmatrix} 0.65 & 0.35 \\ 0.55 & 0.45 \end{bmatrix} \xrightarrow{\times 0.4} \begin{bmatrix} 0.26 & 0.14 \\ 0.55 & 0.45 \end{bmatrix} \xrightarrow{\times 0.6} \begin{bmatrix} 0.26 & 0.24 \\ 0.33 & 0.23 \end{bmatrix}$$

The value of £(y) tells us which rom to select in the column y of the P matrix. The convention for our class is that these numbers are ordered in the same way that they are specified in the support. Problem 4. Consider a DMC whose  $\mathcal{X} = \{1, 2, 3\}, \mathcal{Y} = \{1, 2, 3\}, \text{ and } \mathbf{Q} = \begin{bmatrix} 0.5 & 0.2 & 0.3 \\ 0.5 & 0.2 & 0.3 \\ 0.3 & 0.4 & 0.3 \\ 0.2 & 0.2 & 0.6 \end{bmatrix}$ .

Suppose the input probability vector is  $\mathbf{p} = [0.2, 0.4, 0.4]$ .

(a) Find the joint pmf matrix **P**.

We can get the **P** matrix by scaling each row of the **Q** matrix using the corresponding input probability **p**(**x**).

Method 1:

(b) Find the row vector  $\underline{\mathbf{q}}$  which contains the pmf of the channel output Y.

Method 2: 
$$q_{6} = \rho_{2} P = \begin{bmatrix} 0.2 & 0.4 & 0.4 \end{bmatrix} \begin{bmatrix} 0.5 & 0.2 & 0.3 \\ 0.3 & 0.4 & 0.3 \\ 0.2 & 0.2 & 0.6 \end{bmatrix} = \begin{bmatrix} 1 & 2 & 3 \\ 0.28 & 0.42 \end{bmatrix}$$

(c) Find the following probabilities:

(i) 
$$P[X=1] = 0.2$$

(ii) 
$$P[Y=2] = 0.28$$

(iii) 
$$P[X = 1, Y = 2] = 0.04$$

(iv) 
$$P[Y = 2|X = 1] = 0.2$$

(vi) Find the error probability of the naive decoder. Naive Decoder: 2 Naive (v) =>

$$Q = \begin{bmatrix} 0.5 & 0.2 & 0.3 \\ 0.5 & 0.2 & 0.3 \\ 0.3 & 0.4 & 0.3 \\ 0.2 & 0.2 & 0.6 \end{bmatrix} \xrightarrow{\times 0.7} \begin{bmatrix} 0.10 & 0.07 & 0.06 \\ 0.10 & 0.07 & 0.06 \\ 0.12 & 0.16 & 0.12 \\ 0.08 & 0.08 & 0.08 \end{bmatrix} = \begin{bmatrix} 0.10 & 0.07 & 0.06 \\ 0.08 & 0.08 & 0.08 \end{bmatrix}$$

For naive decoder, look at each column of P and select (circle) the element whose corresponding x value is the <u>same</u> as y in that column.

$$P(E) = 1 - P(C) = 1 - 0.5 = 0.5$$

(vii) Find the error probability of the (DIY) decoder  $\hat{x}(y) = 4 - y$ .  $\Rightarrow$  Decoder table:  $\frac{y}{1}$   $\Rightarrow$  2

$$Q = \begin{bmatrix} 0.5 & 0.2 & 0.3 \\ 0.5 & 0.2 & 0.3 \\ 0.3 & 0.4 & 0.3 \\ 0.2 & 0.2 & 0.6 \end{bmatrix} \xrightarrow{\times 0.7} \begin{bmatrix} 0.10 & 0.07 & 0.06 \\ 0.12 & 0.16 & 0.12 \\ 0.08 & 0.08 & 0.24 \end{bmatrix} = \begin{bmatrix} 0.10 & 0.07 & 0.06 \\ 0.08 & 0.08 & 0.24 \end{bmatrix}$$

For DIY decoder, look at each column of P and select the element whose corresponding x value is the <u>same</u> as  $\hat{x}(y)$  in the decoder table.

NUL SOH STX ETX EOT ENQ ACK BEL BS HT LF CR SO

$$P(C) = 0.08 + 0.16 + 0.06 = 0.30$$

**Problem 5.** A DMC has  $\mathcal{X} = \{0,1\}$  and  $\mathcal{Y} = \{1,2,3\}$ . The following decoding table is used to decode the channel output.

The ASCII Coded Character Set

 $\begin{array}{c|cccc} y & \hat{x}(y) \\ \hline 1 & 0 \\ 2 & 1 \\ 3 & 0 \\ \end{array}$ 

Change all '2' to '1'. Change all '1' and '3' to '0'

Suppose the channel output string is 212213221133122122132.

(a) Find the corresponding decoded string.

## 101100110000011011001

(b) Suppose the channel input string is produced from an ASCII source encoder by the command dec2bin(SourceString,7) in MATLAB. Assume that there is no channel decoding error. Find the corresponding source string.

(There is no need to actually use MATLAB to so we this problem.

(There is no need to actually use MATLAB to solve this problem.

Hint: See pages 7–9 of the slides for Chapter 2.)

## Extra Questions

Here are some optional questions for those who want more practice.

Problem 6. Optimal code lengths that require one bit above entropy: The source coding theorem says that the Huffman code for a random variable X has an expected length strictly less than H(X) + 1. Give an example of a random variable for which the expected length of the Huffman code (without any source extension) is very close to H(X) + 1.

We want to come up with some simple example. Therefore, we shall start by considering the Bernoulli RV X which has only two possible values:

Because there are only two possible values, the pairing in the Auffman coding process must be between these two values:

se 
$$p(n)$$
 c(se)  $\ell(n)$ 
0 1- $\ell_1$ 
1  $\ell_1$ 
1 1 1

Therefore, Huffman coding (without extension) always needs 1 bit per symb≬l.

Now, we want the expected length to be 
$$2H(X)+1$$

=1

So, we must have  $H(X) \ge 0$ .

By definition,  $H(X) = -\frac{2}{\alpha} p(\alpha) \log_2 p(\alpha) = -\frac{1-\alpha}{\alpha} \log_2 1-\alpha - \frac{1-\alpha}{\alpha} \log_2 1-\alpha$ 

You may recall that in Section 2.4, this expression is called the binary entropy function. The plot of this function is shown in the lecture notes; it is sketches below:



Also note that we don't want to have p<sub>4</sub> = 0 or p<sub>5</sub> = 1 because they would make our RV X degenerated (deterministic). For degenerated RV, we don't have to waste any bit to convey its value. (The value is already pre-determined.) For example, if p₁ = 1, we know that P[X = 1] = 1 and hence X≤1 all the time. There is no uncertainty. Anyone can guess the value of X with 100% accuracy by simply guessing the value 1 every time. Alternatively, we may think of the situation here as sending the empty string ( $\epsilon$ ) to the receiver.

So, 
$$\mathbb{E}[\mathcal{L}(x)] = \mathbb{E}[0] = 0$$
.  
Because  $H(x)$  is also 0, we have  $\mathbb{E}[\mathcal{L}(x)] = H(x)$  and not  $H(x) + 1$ .

More formally we can take  $\lim_{n\to 0}$  or  $\lim_{n\to 0}$  on the function and show that  $H(x)\to 0$ .

$$\lim_{\alpha \to 0^+} \alpha \ln \alpha = \lim_{\alpha \to 0^+} \frac{\overline{R}}{-\frac{1}{\alpha c}} = \lim_{\alpha \to 0^+} (-\infty) = 0$$

$$\lim_{\alpha \to 0^+} \alpha \ln \alpha = \lim_{\alpha \to 0^+} \frac{\overline{R}}{-\frac{1}{\alpha c}} = \lim_{\alpha \to 0^+} (-\infty) = 0$$

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$$\lim_{\alpha \to 0^+} \alpha \ln \alpha = \lim_{\alpha \to 0^+} \frac{\overline{R}}{-\frac{1}{\alpha c}} = \lim_{\alpha \to 0^+} (-\infty) = 0$$

So, if p is close to 1 or 0, the entropy will be almost 0. But we still need E[L(x)] = 1 bit to send x. so, the expected length will be very close to HLXI+1.

**Problem 7.** Continue from the Example 2.40 in the lecture and Problem 1. A memoryless source emits two possible message Y(es) and N(o) with probability 0.9 and 0.1, respectively.

- (a) Use MATLAB to find the expected codeword length **per (source) symbol** of the Huffman binary code for the fourth-order extensions of this source.
- (b) Use MATLAB to plot the expected codeword length **per** (source) symbol of the Huffman binary code for the *n*th-order extensions of this source for n = 1, 2, ..., 8.

We first need to list all the probabilities for the (extended) source string (vector/block.)

We then put together these value in a pmf vector using for-loop.

Idea 2: Use MATLAB's "kron" command

Defn. 
$$kron(A,B) = \begin{cases} a_{11}B & a_{12}B & \cdots \\ a_{24}B & a_{22}B & \cdots \\ \vdots & \vdots & \ddots \end{cases}$$

Observe that the pmf used for the nth-order source extension is the Kronecker product of the pmf used for the 1st-order extension

and the pmf used for the (n-1)th order extension

(a) IE[ $l(x_1x_2x_3x_4)$ ] = 1.9702 bits per 4 source symbols

⇒ Ly= 0.4925 bits per source symbol.

